Data-Driven Structured Robust Control of Linear Systems

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Noisy data is *collected* from system observations Control channels are *limited* by network structure



Choose u = Kx to regulate performance of system

Data-Driven	Synthesize K directly from data, no sysid		
Structured	Require that K is in given a subspace		
Robust Control	Regulate worst-case H_2 performance		

of Linear Systems

Structured Control

Set-Membership Data-Driven Control

Merge them together!

Structured Control

Problem: K constrained to a known subspace S

Example 1: sparse control

$$K : \begin{bmatrix} \bullet & \bullet & 0 & 0 \\ 0 & \bullet & \bullet & 0 \\ 0 & 0 & \bullet & \bullet \end{bmatrix}$$
(1)

Example 2: sharing control $(\mathbf{1}^{ op} u = 0)$

$$\forall x : \mathbf{1}^{\top}(Kx) = 0 \implies \mathbf{1}^{\top}K = 0$$
(2)

Structured control is generically NP-hard ¹

Quadratic Invariance: property of graph, dynamic control ² We will use a convex (but conservative) scheme³

¹V. Blondel and J. N. Tsitsiklis, "NP-hardness of some linear control design problems," SIAM J. Control Optim., vol. 35, no. 6, pp. 2118–2127, 1997 ²L. Lessard and S. Lall, "Quadratic invariance is necessary and sufficient for convexity," in Proceedings of the 2011 American Control Conference, 2011, pp. 5360–5362.

³Ferrante, Francesco, Fabrizio Dabbene, and Chiara Ravazzi. "On the design of structured stabilizers for LTI systems." IEEE Control Systems Letters 4.2 (2019): 289-294.

Robust Design

 H_2 norm $< \gamma$ under u = Kx if program is feasible⁴:

$$\begin{aligned}
& \text{find}_{P,Q,R,L} \quad \begin{bmatrix} P - EE^{\top} & AR + BL \\ (AR + BL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0 & (3a) \\
& \begin{bmatrix} Q & CR + DL \\ (CR + DL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0 & (3b) \\
& \text{Tr}(Q) \leq \gamma^2 & (3c) \\
& P \in \mathbb{S}^n, \ Q \in \mathbb{S}^q, \ R \in \mathbb{R}^{n \times n}, \ L \in \mathbb{R}^{m \times n}. & (3d)
\end{aligned}$$

<u>If feasible, then R is nonsingular with valid $K = LR^{-1}$ </u>

⁴Thm. 5: De Oliveira, Mauricio C., José C. Geromel, and Jacques Bernussou. "Extended H_2 and H_{∞} norm characterizations and controller parametrizations for discrete-time systems." International journal of control 75.9 (2002): 666-679. We desire $K = LR^{-1} \in S$, but this is generally nonconvex

top block:
$$\begin{bmatrix} P - EE^{\top} & AR + BL \\ (AR + BL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0$$
(4)

Only *R* and *L* are sparsity constrained, *P* can be dense Standard approach: $L \in S$, *R* is *diagonal* \implies $K \in S$

Structured Control with Reduced Conservatism

R diagonal is too strict, can be loosened ⁵ Given basis $\{S_{\ell}\}_{\ell=1}^{k}$ for *S* with $(S = \text{span}(\{S_{\ell}\}))$

Define representation S and convex set $\Upsilon(S)$ as

$$S := \begin{bmatrix} S_1 & S_2 & \dots & S_k \end{bmatrix}$$
(5)

 $\Upsilon(S) := \{ Q \in \mathbb{R}^{n \times n} \mid \exists \Lambda \in \mathbb{S}^k : S(I_k \otimes Q) = S(\Lambda \otimes I_n) \}$ (6)

Subspace-compatible controller

$$L \in S, R \in \Upsilon(S), |R| \neq 0 \implies LR^{-1} \in S$$
 (7)

⁵Ferrante, Francesco, Fabrizio Dabbene, and Chiara Ravazzi. "On the design of structured stabilizers for LTI systems." IEEE Control Systems Letters 4.2 (2019): 289-294.

Convex-Relaxed Structured Robust Design

 H_2 norm $< \gamma$ for u = Kx with $K \in S$ if program is feasible

$$\begin{array}{l} \operatorname{find}_{P,Q,R,L} & \begin{bmatrix} P - EE^{\top} & AR + BL \\ (AR + BL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0 \quad (8a) \\ & \begin{bmatrix} Q & CR + DL \\ (CR + DL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0 \quad (8b) \\ & \operatorname{Tr}(Q) \leq \gamma^2 \quad (8c) \\ & P \in \mathbb{S}^n, \ Q \in \mathbb{S}^q, \ R \in \Upsilon(S), \ L \in \mathcal{S}. \quad (8d) \end{array}$$

If feasible, then R is nonsingular with valid $K = LR^{-1} \in S$ Still requires *known* plant (A, B)

Structured Data-Driven Control

Virtual Reference Feedback Tuning (first methods) **Set-Membership** (this talk)

- (Data-consistent plants) \subseteq (K-Stabilized plants)
- Certificates: Farkas, Interval, S-Lemma (QMI), SOS

Behavioral

- Description of all signal relations
- Parameterize and pick out best system trajectory (DeepC)

Others: Koopman, Gaussian Processes

Data Collected

Data (x, u) collected as a *T*-length trajectory:

Unknown (but bounded) observation noise process w

$$\mathbf{W} := [w(0) \quad w(1) \quad \dots \quad w(T-1)] \in \mathbb{R}^{n \times T}$$

(9)

Ground truth (A_*, B_*) obeys

$$\mathbf{X}_{+} = A_* \mathbf{X}_{-} + B_* \mathbf{U} + \mathbf{W}$$
(10)

Noise Description

Noise sequence w bounded inside matrix ellipsoid ⁶:

$$\begin{bmatrix} I \\ \mathbf{W}^{\top} \end{bmatrix} \begin{bmatrix} \Phi_{11} & \Phi_{12} \\ \Phi_{12}^{\top} & \Phi_{22} \end{bmatrix} \begin{bmatrix} I \\ \mathbf{W}^{\top} \end{bmatrix}^{\top} \succeq 0.$$
(11)

for $\Phi \in \mathbb{S}^{n+T}$ with $\Phi_{11} \in \mathbb{S}^n_+$ and $-\Phi_{22} \in \mathbb{S}^T_{++}$

Process noise bound $\forall t : ||w(t)||_2 \le \epsilon$ (overapproximation)

$$\Phi = \begin{bmatrix} \mathcal{T} \epsilon I_n & \mathbf{0} \\ \mathbf{0} & -I_T \end{bmatrix}.$$
 (12)

⁶Van Waarde, Henk J., et al. "Quadratic matrix inequalities with applications to data-based control." SIAM Journal on Control and Optimization 61.4 (2023): 2251-2281.

Data-Consistency Set

Data-bound-characterizing matrix $\boldsymbol{\Psi}$

$$\Psi := \begin{bmatrix} I & \mathbf{X}_{+} \\ \mathbf{0} & -\mathbf{X}_{-} \\ \mathbf{0} & -\mathbf{U} \end{bmatrix}^{\top} \Phi \begin{bmatrix} I & \mathbf{X}_{+} \\ \mathbf{0} & -\mathbf{X}_{-} \\ \mathbf{0} & -\mathbf{U} \end{bmatrix}.$$
(13)

Set of data-consistent plants (A, B):

$$\Sigma_{\mathcal{D}} = \left\{ (A, B) \mid \begin{bmatrix} I \\ A^{\top} \\ B^{\top} \end{bmatrix}^{\top} \Psi \begin{bmatrix} I \\ A^{\top} \\ B^{\top} \end{bmatrix} \succeq 0 \right\}.$$
(14)

Equivalent statements:

$$\begin{bmatrix} P - EE^{\top} & AR + BL \\ (AR + BL)^{\top} & R + R^{\top} - P \end{bmatrix} \succ 0$$
$$\begin{bmatrix} I \\ A^{\top} \\ B^{\top} \end{bmatrix}^{\top} \begin{bmatrix} P - EE^{\top} & \mathbf{0} \\ \mathbf{0} & -\begin{bmatrix} R \\ L \end{bmatrix} (R + R^{\top} - P)^{-1} \begin{bmatrix} R \\ L \end{bmatrix}^{\top} \begin{bmatrix} I \\ A^{\top} \\ B^{\top} \end{bmatrix} \succ 0$$

Must be enforced $\forall (A, B) \in \Sigma_D$ with common (P, R, L)

Matrix S-Lemma⁷: exist constants $\alpha \ge 0$, $\beta > 0$ such that:

$$\begin{bmatrix} P - EE^{\top} - \beta I & \mathbf{0} \\ \mathbf{0} & - \begin{bmatrix} R \\ L \end{bmatrix} (R + R^{\top} - P)^{-1} \begin{bmatrix} R \\ L \end{bmatrix}^{\top} \end{bmatrix} - \alpha \Psi \succeq \mathbf{0}.$$

Nonconservative (over 1 quadratic constraint)

⁷Van Waarde, Henk J., M. Kanat Camlibel, and Mehran Mesbahi. "From noisy data to feedback controllers: Nonconservative design via a matrix S-lemma." IEEE Transactions on Automatic Control 67.1 (2020): 162-175.

Expand by Schur Complement

$$\begin{bmatrix} P - EE^{\top} - \beta I & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \mathbf{0} & R^{\top} & L^{\top} & R + R^{\top} - P \end{bmatrix} - \alpha \begin{bmatrix} \Psi & \mathbf{0} \\ \mathbf{0} & \mathbf{0} \end{bmatrix} \succ \mathbf{0}$$

Now is an LMI (convex expression) in (R, L, P)

$$\begin{array}{ll} \inf_{P,R,L,\alpha,\beta,\gamma} & \gamma & \text{subject to:} \\ \hline P - EE^\top - \beta I & \mathbf{0} & \mathbf{0} & \mathbf{0} \\ \hline \mathbf{0} & \mathbf{0} & \mathbf{0} & R \\ \hline \mathbf{0} & \mathbf{0} & \mathbf{0} & L \\ \hline \mathbf{0} & R^\top & L^\top & R + R^\top - P \\ \hline \mathbf{0} & R^\top & L^\top & R + R^\top - P \\ \hline \begin{bmatrix} Q & CR + DL \\ (CR + DL)^\top & R + R^\top - P \\ \end{bmatrix} \succeq \mathbf{0} & (15b) \\ \hline \operatorname{Tr}(Q) \leq \gamma^2 & (15c) \\ \alpha \geq \mathbf{0}, \beta > \mathbf{0}, \gamma \geq \mathbf{0} & (15d) \\ P \in \mathbb{S}^n, \ Q \in \mathbb{S}^q, \ R \in \Upsilon(S), \ L \in \mathcal{S} & (15e) \\ \end{array}$$

Examples

Three-State Two-Input System

Ground truth of

$$A_* = \begin{bmatrix} -0.4095 & 0.4036 & -0.0874 \\ 0.5154 & -0.0815 & 0.1069 \\ 1.6715 & 0.7718 & -0.3376 \end{bmatrix}$$
$$B_* = \begin{bmatrix} 0 & 0 \\ -0.6359 & -0.1098 \\ -0.0325 & 2.2795 \end{bmatrix}$$

H₂-suboptimal control with

$$C = \begin{bmatrix} I_3 \\ \mathbf{0}_{2\times 3} \end{bmatrix}, \qquad D = \begin{bmatrix} \mathbf{0}_{3\times 2} \\ I_2 \end{bmatrix}, \qquad E = I_3. \tag{17}$$

(16)

Sparse Control Task

Controller structure:

$$\mathcal{K} \in \begin{bmatrix} \bullet & \bullet & 0 \\ 0 & \bullet & \bullet \end{bmatrix} \qquad \Upsilon(S) : \begin{bmatrix} R_{11} & R_{12} & 0 \\ 0 & R_{22} & 0 \\ 0 & R_{32} & R_{33} \end{bmatrix} \qquad (18)$$

Control tasks:

- 1. Unstructured (dense, comparison)
- 2. P = R diagonal, $L \in S$
- 3. *P* dense, *R* diagonal, $L \in S$
- 4. *P* dense, $R \in \Upsilon(S)$, $L \in S$ (our scheme)

Closed-loop H_2 upper-bounds v.s. ϵ with T = 20

Design	(A_*,B_*)	$\epsilon = 0.05$	$\epsilon = 0.1$	$\epsilon = 0.15$
1 (Unstructured)	2.1537	2.3448	3.0939	4.5757
2 (P = R diag.)	3.5658	4.6619	7.4193	Infeasible
3 (<i>R</i> diag.)	3.0089	3.5997	4.9506	9.1999
$4\;(R\in\Upsilon(S))$	2.9794	3.5495	4.6806	8.9710

Closed-loop H_2 upper-bounds v.s. T with $\epsilon = 0.1$

Design	(A_*,B_*)	<i>T</i> = 6	T = 10	<i>T</i> = 20
1 (Unstructured)	2.1537	2.9911	2.8156	3.0939
2 (P = R diag.)	3.5658	6.3386	7.0963	7.4193
3 (<i>R</i> diag.)	3.0089	4.5545	4.5044	4.9506
$4\;(R\in\Upsilon(S))$	2.9794	4.4036	4.4323	4.6806

Non-monotonicity due to overapproximation



Structured H_2 regulation based on data

Convex $\Upsilon(S)$ description of S constraint

Matrix S-Lemma for robust certification

Extensions: LPV synthesis, determining optimal structures

Thanks!